Part-II Parametric Signal Modeling and Linear Prediction Theory2. Discrete Wiener Filtering

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Acknowledgment: ENEE630 slides were based on class notes developed by Profs. K.J. Ray Liu and Min Wu. The LaTeX slides were made by Prof. Min Wu and Mr. Wei-Hong Chuang.

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Preliminaries

[Readings: Haykin's 4th Ed. Chapter 2, Hayes Chapter 7]

• Why prefer FIR filters over IIR?

 \Rightarrow FIR is inherently stable.

- Why consider complex signals?
 - Baseband representation is complex valued for narrow-band messages modulated at a carrier frequency.
 - Corresponding filters are also in complex form.
- $u[n] = u_I[n] + j u_Q[n]$
- $u_l[n]$: in-phase component • $u_Q[n]$: quadrature component the two parts can be amplitude



the two parts can be amplitude modulated by $\cos 2\pi f_c t$ and $\sin 2\pi f_c t$.

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(1) General Problem

(Ref: Hayes §7.1)



Want to process x[n] to <u>minimize</u> the difference between the estimate and the desired signal in some sense:

A major class of estimation (for simplicity & analytic tractability) is to use linear combinations of x[n] (i.e. via linear filter).

When x[n] and d[n] are from two <u>w.s.s.</u> random processes, we often choose to minimize the mean-square error as the performance index.

$$\min_{\underline{w}} J \triangleq \mathbb{E}\left[|e[n]|^2\right] = \mathbb{E}\left[|d[n] - \hat{d}[n]|^2\right]$$

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(2) Categories of Problems under the General Setup

- Filtering
- Smoothing
- In Prediction
- Occonvolution

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Wiener Problems: Filtering & Smoothing

- Filtering
 - The classic problem considered by Wiener
 - x[n] is a noisy version of d[n]: x[n] = d[n] + v[n]
 - The goal is to estimate the true d[n] using a causal filter (i.e., from the current and post values of x[n])
 - The causal requirement allows for filtering on the fly
- Smoothing
 - Similar to the filtering problem, except the filter is allowed to be non-causal (i.e., all the *x*[*n*] data is available)

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Wiener Problems: Prediction & Deconvolution

- Prediction
 - The causal filtering problem with d[n] = x[n+1],
 - i.e., the Wiener filter becomes a linear predictor to predict x[n+1] in terms of the linear combination of the previous value $x[n], x[n-1], \ldots$
- Deconvolution
 - To estimate d[n] from its filtered (and noisy) version x[n] = d[n] * g[n] + v[n]
 - If g[n] is also unknown ⇒ blind deconvolution.
 We may iteratively solve for both unknowns

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FIR Wiener Filter for w.s.s. processes

Design an FIR Wiener filter for jointly w.s.s. processes $\{x[n]\}$ and $\{d[n]\}$: $W(z) = \sum_{k=0}^{M-1} a_k z^{-k}$ (where a_k can be complex valued) $\hat{d}[n] = \sum_{k=0}^{M-1} a_k x[n-k] = \underline{a}^T \underline{x}[n]$ (in vector form) $\Rightarrow e[n] = d[n] - \hat{d}[n] = d[n] - \sum_{k=0}^{M-1} \underbrace{a_k x[n-k]}_{\hat{d}[n] = \underline{a}^T \underline{x}[n]}$

By summation-of-scalar: $J = E[|e(n)|^{2}] = E[e(n) e^{x}(n)]$ $= E[|d(n)|^{2}] - E[d(n) \sum_{k=0}^{n} a_{k} x^{*}(n-k)] - E[d^{*}(n) \sum_{k=0}^{n-1} a_{k} x(n-k)] + E[\sum_{k=0}^{n-1} a_{k} a_{k} a_{k}^{*} x(n-k) x(n-k)]$ $= E[(d(n)|^{2}] - \sum_{k=0}^{n-1} a_{k}^{*} E[d(n) x^{*}(n-k)] - \sum_{k=0}^{n-1} a_{k} E[d^{*}(n) x(n-k)] + \sum_{k=0}^{n-1} a_{k} a_{k}^{*} E[x(n+k) x(n-k)]$ $= E[(d(n)|^{2}] - \sum_{k=0}^{n-1} a_{k}^{*} E[d(n) x^{*}(n-k)] - \sum_{k=0}^{n-1} a_{k} E[d^{*}(n) x(n-k)] + \sum_{k=0}^{n-1} a_{k} a_{k}^{*} E[x(n+k) x(n-k)]$

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FIR Wiener Filter for w.s.s. processes

In matrix-vector form:

$$J = \mathbb{E}\left[|d[n]|^{2}\right] - \underline{a}^{H}\underline{p}^{*} - \underline{p}^{T}\underline{a} + \underline{a}^{H}\mathbf{R}\underline{a}$$

where $\underline{x}[n] = \begin{bmatrix} x[n] \\ x[n-1] \\ \vdots \\ x[n-M+1] \end{bmatrix}$, $\underline{p} = \begin{bmatrix} \mathbb{E}[x[n]d^{*}[n]] \\ \vdots \\ \mathbb{E}[x[n-M+1]d^{*}[n]] \end{bmatrix}$,
 $\underline{a} = \begin{bmatrix} a_{0} \\ \vdots \\ a_{M-1} \end{bmatrix}$.

- $\mathbb{E}\left[|d[n]|^2\right]$: σ^2 for zero-mean random process
- $\underline{a}^{H}\mathbf{R}\underline{a}$: represent $\mathbb{E}\left[\underline{a}^{T}\underline{x}[n]\underline{x}^{H}[n]\underline{a}^{*}\right] = \underline{a}^{T}\mathbf{R}\underline{a}^{*}$

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Perfect Square

() If **R** is positive definite, \mathbf{R}^{-1} exists and is positive definite.

$$(\mathbf{R}\underline{a}^* - \underline{p})^H \mathbf{R}^{-1} (\mathbf{R}\underline{a}^* - \underline{p}) = (\underline{a}^T \mathbf{R}^H - \underline{p}^H) (\underline{a}^* - \mathbf{R}^{-1}\underline{p}) = \underline{a}^T \mathbf{R}^H \underline{a}^* - \underline{p}^H \underline{a}^* - \underline{a}^T \underbrace{\mathbf{R}^H \mathbf{R}^{-1}}_{=\mathbb{I}} \underline{p} + \underline{p}^H \mathbf{R}^{-1} \underline{p}$$

Thus we can write $J(\underline{a})$ in the form of perfect square:

$$J(\underline{a}) = \underbrace{\mathbb{E}\left[|d[n]|^2\right] - \underline{p}^H \mathbf{R}^{-1} \underline{p}}_{\text{Not a function of }\underline{a}; \text{ Represent } J_{\min}.} + \underbrace{(\mathbf{R}\underline{a}^* - \underline{p})^H \mathbf{R}^{-1} (\mathbf{R}\underline{a}^* - \underline{p})}_{>0 \text{ except being zero if } \mathbf{R}\underline{a}^* - \underline{p}=0}$$

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Perfect Square

 $J(\underline{a})$ represents the error performance surface: convex and has unique minimum at $\mathbf{R}\underline{a}^* = \underline{p}$



Ja

Thus the necessary and sufficient condition for determining the optimal linear estimator (linear filter) that minimizes MSE is

$$\mathbf{R}\underline{a}^* - \underline{p} = 0 \Rightarrow \mathbf{R}\underline{a}^* = \underline{p}$$

This equation is known as the **Normal Equation**. A FIR filter with such coefficients is called a **FIR Wiener filter**.

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Perfect Square

$$\mathbf{R}\underline{a}^* = \underline{p} \quad \therefore \underline{a}^*_{opt} = \mathbf{R}^{-1}\underline{p} \text{ if } \mathbf{R} \text{ is not singular}$$
(which often holds due to noise)

When $\{x[n]\}$ and $\{d[n]\}$ are jointly w.s.s. (i.e., crosscorrelation depends only on time difference)



This is also known as the Wiener-Hopf equation (the discrete-time counterpart of the continuous Wiener-Hopf integral equations)

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Principle of Orthogonality

Note: to minimize a real-valued func. $f(z, z^*)$ that's analytic (differentiable everywhere) in z and z^* , set the derivative of f w.r.t. either z or z^* to zero.

• Necessary condition for minimum $J(\underline{a})$: (nece.&suff. for convex J) $\frac{\partial}{\partial a_k^*}J = 0$ for $k = 0, 1, \dots, M - 1$. $\Rightarrow \frac{\partial}{\partial a_k^*}\mathbb{E}\left[e[n]e^*[n]\right] = \mathbb{E}\left[e[n]\frac{\partial}{\partial a_k^*}(d^*[n] - \sum_{j=0}^{M-1} a_j^*x^*[n-j])\right]$ $= \mathbb{E}\left[e[n] \cdot (-x^*[n-k])\right] = 0$

Principal of Orthogonality

$$\mathbb{E}\left[e_{\text{opt}}[n]x^*[n-k]\right] = 0 \text{ for } k = 0, \dots, M-1.$$

The optimal error signal e[n] and each of the M samples of x[n] that participated in the filtering are statistically uncorrelated (i.e., orthogonal in a statistical sense)

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Principle of Orthogonality: Geometric View



 $\begin{array}{l} \mbox{Analogy:} \\ \mbox{r.v.} \ \Rightarrow \mbox{vector;} \\ \mbox{E}(XY) \ \Rightarrow \mbox{inner product of vectors} \end{array}$

⇒ The optimal $\hat{d}[n]$ is the projection of d[n] onto the subspace spanned by $\{x[n], ..., x[n - M + 1]\}$ in a statistical sense.

The vector form: $\mathbb{E}\left[\underline{x}[n]e_{opt}^{*}[n]\right] = \underline{0}.$

This is true for any linear combination of $\underline{x}[n]$ and for FIR & IIR:

$$\mathbb{E}\left[\hat{d}_{\rm opt}[n]e_{\rm opt}[n]\right] = 0$$

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Minimum Mean Square Error

Recall the perfect square form of *J*:

$$J(\underline{a}) = \underbrace{\mathbb{E}\left[|d[n]|^2\right] - \underline{p}^H \mathbf{R}^{-1} \underline{p}}_{d} + \underbrace{(\mathbf{R}\underline{a}^* - \underline{p})^H \mathbf{R}^{-1} (\mathbf{R}\underline{a}^* - \underline{p})}_{d} + \underbrace{(\mathbf{R}\underline{a}^* - \underline{p})^H \mathbf{R}^{-1} (\mathbf{R}\underline{a}^* - \underline{p})}_{d} + \underbrace{(\mathbf{R}\underline{a}^* - \underline{p})^H \mathbf{R}^{-1} (\mathbf{R}\underline{a}^* - \underline{p})}_{d} + \underbrace{(\mathbf{R}\underline{a}^* - \underline{p})^H \mathbf{R}^{-1} \underline{p}}_{d}$$
Also recall $d[n] = \hat{d}_{opt}[n] + e_{opt}[n]$. Since $\hat{d}_{opt}[n]$ and $e_{opt}[n]$ are uncorrelated by the principle of orthogonality, the variance is $\sigma_d^2 = \operatorname{Var}(\hat{d}_{opt}[n]) + J_{min}$
 $\therefore \operatorname{Var}(\hat{d}_{opt}[n]) = \underline{p}^H \mathbf{R}^{-1} \underline{p}$

$$= \underline{a}_0^H \underline{p}^* = \underline{p}^H \underline{a}_o^* = \underline{p}^T \underline{a}_o \quad \text{real and scalar}$$

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Example and Exercise



- What kind of process is {x[n]}?
- What is the correlation matrix of the channel output?
- What is the cross-correlation vector?

•
$$w_1 = ?$$
 $w_2 = ?$ $J_{\min} = ?$

Detailed Derivations

Another Perspective (in terms of the gradient)

Theorem: If $f(\underline{z}, \underline{z}^*)$ is a **real-valued** function of complex vectors \underline{z} and \underline{z}^* , then the vector pointing in the direction of the maximum rate of the change of f is $\nabla_{\underline{z}^*} f(\underline{z}, \underline{z}^*)$, which is a vector of the derivative of f() w.r.t. each entry in the vector \underline{z}^* .

Corollary: Stationary points of $f(\underline{z}, \underline{z}^*)$ are the solutions to $\nabla_{\underline{z}^*} f(\underline{z}, \underline{z}^*) = 0$.

Complex gradient of a complex function: $\begin{array}{c|c}
\underline{a}^{H}\underline{z} & \underline{z}^{H}\underline{a} & \underline{z}^{H}A\underline{z} \\
\hline & & \\
\nabla_{\underline{z}} & \underline{a}^{*} & 0 & A^{T}\underline{z}^{*} = (A\underline{z})^{*} \\
\hline & & \\
\nabla_{z^{*}} & 0 & a & Az
\end{array}$

Using the above table, we have $\bigtriangledown_{\underline{a}^*} J = -\underline{p}^* + \mathbf{R}^T \underline{a}$.

For optimal solution: $\bigtriangledown_{\underline{a}^*} J = \frac{\partial}{\partial \underline{a}^*} J = 0$ $\Rightarrow \mathbf{R}^T \underline{a} = \underline{p}^*$, or $\mathbf{R} \underline{a}^* = \underline{p}$, the Normal Equation. $\therefore \underline{a}^*_{opt} = \mathbf{R}^{-1} \underline{p}$

(Review on matrix & optimization: Hayes 2.3; Haykins(4th) Appendix A,B,C)

Review: differentiating complex functions and vectors

(1) Differentiableat do Need to converge
$$\begin{array}{c} (1) \quad \underbrace{f(\mathcal{F}_{0}(d)) - f(\mathcal{F}_{0})}{\mathcal{F}_{0}(d)} \quad exist \implies in all directions \\ \mathcal{F}_{0} = \mathcal{F}_{0} \quad \mathcal{F}_{0} \implies in all directions \\ \mathcal{F}_{0} = \mathcal{F}_{0} \quad \mathcal{F}_{0} \implies in all directions \\ \mathcal{F}_{0} = \mathcal{F}_{0} \implies is analytic (i.e. differentiable every where) on Negion D if \\ f(\mathcal{F}) = W(X, Y) + i \mathcal{V}(X, Y) \quad is conthinuous and satisfy (auchy - Riemann \\ Ondition \qquad \underbrace{\mathcal{F}_{0}}_{\mathcal{F}_{0}} = \underbrace{\mathcal{F}_{0}}_{\mathcal{F}_{0}} \stackrel{\mathcal{F}_{0}}{\mathcal{F}_{0}} = -\frac{\mathcal{F}_{0}}{\mathcal{F}_{0}} \\ \end{array}$$

$$\begin{array}{c} (2) \quad e.g. \quad f_{i}(\mathcal{F}) = \mathcal{F}_{0}^{*} = |\mathcal{F}|^{2} = (\chi^{2} + \chi^{2}) + i \times 0 \\ \qquad f_{1}(\mathcal{F}) = \mathcal{F}_{0}^{*} = \chi - i \chi \\ \implies \text{DOES NOT S atisfy Cauchy - Riemann.} \end{array}$$

Review: differentiating complex functions and vectors

unlike
the ← Note:
$$f(\delta) = |\delta|^{\perp}$$
 has unique minimum at $\beta = 0$, but not
term ← Note: $f(\delta) = |\delta|^{\perp}$ has unique minimum at $\beta = 0$, but not
totale
differentiable-from complex analysis (any func. that depends on β^{+} is not differentiable)
optimis.
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Differentiating complex functions: More details



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